

# **RHFL4913 ADJUSTABLE**

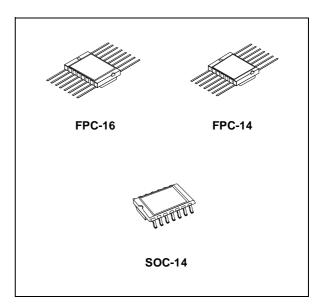
## **RAD-HARD ADJUSTABLE POSITIVE VOLTAGE REGULATOR**

- 1-3 AMPERE LOW DROPOUT VOLTAGE
- EMBEDDED OVERTEMPERATURE, OVERCURRENT PROTECTIONS
- ADJUSTABLE OVERCURRENT LIMITATION
- OUTPUT OVERLOAD MONITORING/ . SIGNALLING
- ADJUSTABLE OUTPUT VOLTAGE
- INHIBIT (ON/OFF) TTL COMPATIBLE CONTROL
- PROGRAMMABLE OUTPUT SHORT CIRCUIT CURRENT
- REMOTE SENSING OPERATION
- RADHARD: TESTED UP TO 300krad IN MIL 1019.5 AND LOW DOSE RATE CONDITIONS
- HEAVY IONS SEL, SEU IMMUNE. SUSTAINS 2x10<sup>14</sup> proton/cm<sup>2</sup>, AND 2x10<sup>14</sup> neutron/cm<sup>2</sup>

### DESCRIPTION

The RHFL4913 ADJUSTABLE is a high performance Rad Hard Adjustable Positive Voltage Regulator.

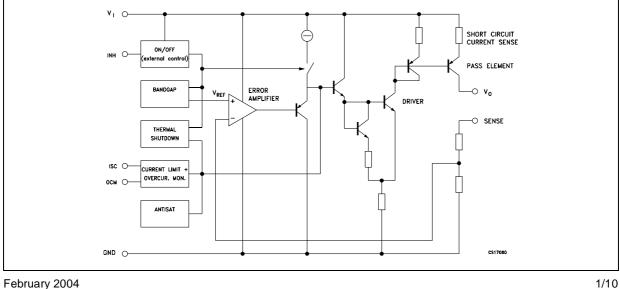
Available into various hermetic ceramic packages, it is specifically intended for Space and harsh



radiation environments. Input supply range is from 3 to 12 volts.

The RHFL4913 ADJUSTABLE is QmI-V Qualified, DSCC Smd is 596202524.

### **BLOCK DIAGRAM**



### ABSOLUTE MAXIMUM RATINGS (Note 1)

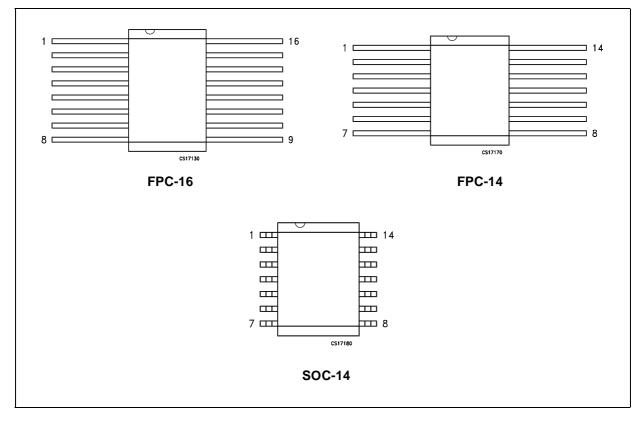
Symbol	Parameter		Value	Unit	
VI	DC Input Voltage, V <sub>I</sub> - V <sub>GROUND</sub>		14	V	
Vo	DC Output Voltage Range		1.3 to 9	V	
1	Output Current	RHFL4913KSOA, KA	1	A	
I <sub>O</sub>		RHFL4913KPA	2		
Р	T <sub>C</sub> = 25°C Power Dissipation KSOA, KA versions (14 Pin)		3	10/	
PD		KPA, versions (16 Pin)	15	W	
T <sub>stg</sub>	Storage Temperature Range		-65 to +150	°C	
T <sub>op</sub>	Operating Junction Temperature Range		-55 to +150	°C	
ESD	Electrostatic Discharge Capal	bility	Class 3		

Note 1: Exceeding maximum ratings may damage the device.

### THERMAL DATA

Symbol	Parameter	FPC-14	SOC-14	FPC-16	Unit
R <sub>thj-case</sub>	Thermal Resistance Junction-case Max	42	42	8.3	°C/W
T <sub>sold</sub>	Maximum soldering Temperature, 10sec.		300		°C

### **CONNECTION DIAGRAM (Top view)**



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### **PIN DESCRIPTION**

PIN N°	FPC-16	FPC-14	SOC-14
V <sub>O</sub>	1, 2, 6, 7	2, 6	2, 6
VI	3, 4, 5	3, 5	3, 5
GND	13	12	12
I <sub>SC</sub>	8	7	7
OCM	10	8	8
INHIBIT	14	13	13
ADJ	15	14	14
NC	9, 11, 12, 16	1, 4, 9, 10, 11	1, 4, 9, 10, 11

### **ORDERING CODES**

FPC-14	FPC-16	SOC-14	SOLDER DIPPING	OUTPUT VOLTAGE
RHFL4913KA-05V	RHFL4913KPA-01V	RHFL4913SOA-03V	GOLD	ADJ
RHFL4913KA-06V	RHFL4913KPA-02V	RHFL4913SOA-04V	SOLDER	ADJ

### PART NUMBER - SMD EQUIVALENCE

ST PART NUMBER	SMD PART NUMBER
RHFL4913KPA-01V	5962F0252401VXC
RHFL4913KPA-02V	5962F0252401VXA
RHFL4913SOA-03V	5962F0252402VYC
RHFL4913SOA-04V	5962F0252402VYA
RHFL4913KA-05V	5962F0252402VZC
RHFL4913KA-06V	5962F0252402VZA

### **ENVIRONMENTAL CHARACTERISTICS**

Parameter Conditions		Typical	Unit
Output Voltage thermal drift	-55°C to 125°C	40	ppm/°C
Output Voltage radiation drift	from 0 krad to 300 krad at 0.55rad/sec	8	ppm/krad
Output Voltage radiation drift	from 0 krad to 300 krad, Mil 1019.5	6	ppm/krad

### **RHFL4913 ADJUSTABLE VERSION**

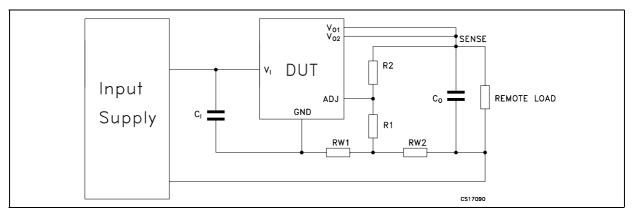
Symbol	Parameter	Test Conditions		Min.	Тур.	Max.	Unit
VI	Operating Input Voltage	$I_0 = 1A$ $T_{J} = -55 \text{ to } 12$	$I_{O} = 1A$ $T_{J} = -55 \text{ to } 125^{\circ}\text{C}$			12	V
V <sub>ADJ</sub>	Adjust Pin Voltage	$V_{I}$ , $V_{O}$ , $I_{O}$ , $P_{D}$ within Max Rating Ii	$V_{I}$ , $V_{O}$ , $I_{O}$ , $P_{D}$ within Max Rating limits			1.27	%
ISHORT	Output Current Limit (*)	Adjustable by mask/external resis	tor		4.5		Α
$\Delta V_{O} / \Delta V_{I}$	Line Regulation	$V_{I} = V_{O}$ +2.5V to 12 V, $I_{O}$ = 5mA				0.35	%
$\Delta V_{O} / \Delta V_{O}$	Load Regulation	$V_{I} = V_{O}$ +2.5V, $I_{O}$ = 5mA to 400 m/	A			0.3	%
		$V_{I} = V_{O}$ +2.5V, $I_{O}$ = 5mA to 1A				0.5	
Z <sub>OUT</sub>	Output Impedance	$I_{O} = 100 \text{ mA DC}$ and 20 mA rms			100		mΩ
۱ <sub>q</sub>	Quiescent Current	$V_{I} = V_{O}$ +2.5V, $I_{O} = 5mA$ On Mo	ode			6	mA
		$V_{I} = V_{O}$ +2.5V, $I_{O} = 30$ mA On Mo	ode			8	
		$V_{I} = V_{O}$ +2.5V, $I_{O} = 300$ mA On Mo	ode			25	
		$V_{I} = V_{O}$ +2.5V, $I_{O} = 1A$ On Mo	ode			60	
		$V_{I} = V_{O}+2V$ , $V_{INH} = 2.4V$ Off Mo			1		
۱ <sub>q</sub>	Quiescent Current	$V_{I} = V_{O}$ +2.5V, $I_{O} = 5$ mA, $T_{J}$ =-55 to 125°C				6	mA
		$V_{I} = V_{O}+2.5V$ , $I_{O} = 30$ mA, $T_{J}=-55$ to $125^{\circ}$ C				14	
		$V_{I} = V_{O}$ +2.5V, $I_{O}$ = 300mA, $T_{J}$ =-58	5 to 125°C			40	
		$V_{I} = V_{O}$ +2.5V, $I_{O} = 1A$ , $T_{J}$ =-55 to $T_{I}$	$V_{I} = V_{O}+2.5V$ , $I_{O} = 1A$ , $T_{J}=-55$ to $125^{\circ}C$			100	
V <sub>d</sub>	Dropout Voltage	$I_{O} = 400 \text{mA}$ $V_{O} = 2.5 \text{ to } 9 \text{ V}, (-55^{\circ}\text{C})$			300	400	V
		$I_0 = 400 \text{mA}$ $V_0 = 2.5 \text{ to } 9 \text{ V}, (25)$	°C)		350	450	
		$I_0 = 1A$ $V_0 = 2.5 \text{ to } 9 \text{ V}, (25)$	°C)			650	
		$I_0 = 2A$ $V_0 = 2.5 \text{ to } 9 \text{ V}, (25)$	°C)		900		
		$I_0 = 400 \text{mA}$ $V_0 = 2.5 \text{ to } 9 \text{ V}, (129)$	5°C)		450	550	
		$I_0 = 1A$ $V_0 = 2.5 \text{ to } 9 \text{ V}, (129)$	5°C)			800	
		$I_{\rm O} = 2A$ $V_{\rm O} = 2.5$ to 9 V, (12)	5°C)		950		
V <sub>INH(ON)</sub>	Inhibit Voltage	$I_{O} = 5$ mA, $T_{J}$ =-55 to 125°C				0.8	V
V <sub>INH(OFF)</sub>	Inhibit Voltage	$I_{O} = 5$ mA, $T_{J}$ =-55 to 125°C		2.4			V
SVR	Supply Voltage Rejection		120Hz	60	70		dB
		÷	33KHz	30	40		
I <sub>SH</sub>	Shutdown Input Current	V <sub>INH</sub> = 5 V			15		μA
V <sub>OCM</sub>	OCM Pin Voltage	Sinked I <sub>OCM</sub> = 10 mA active low			0.38		V
t <sub>PLH</sub>	Inhibit Propagation Delay	$V_{I} = V_{O}+2.5V, V_{INH} = 2.4 V, I_{O} = 4$	400 mA	ON-OFF		20	μS
t <sub>PHL</sub>		-		OFF-ON		100	μS
eN	Output Noise Voltage	B= 10Hz to $100$ KHzI <sub>O</sub> = 5mA to 2	A.		40		$\mu V \text{rms}$

### **ELECTRICAL CHARACTERISTICS** (T<sub>J</sub> = 25°C, V<sub>I</sub> = V<sub>O</sub>+2.5V, C<sub>I</sub> = C<sub>O</sub> = 1 $\mu$ F, unless otherwise specified)

(\*) This value is guaranteed by design. For each application it's strongly recommended to comply with the maximum current limit of the package used.

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### APPLICATION DIAGRAM FOR REMOTE SENSINS OPERATION



### DEVICE DESCRIPTION

The RHFL4913 Adjustable contains a PNP type power element controlled by a signal resulting from amplified comparison between the internal temperature compensated Band-Gap and the fraction of the desired Output Voltage value obtained from an external resistor divider bridge. The device is protected by several functional blocks.

### ADJ pin

The Load output voltage feed-back comes from an external divider resistor bridge middle point connected to ADJ pin (allowing all possible output voltage settings as per User's desire) established between Load terminals.

### **INHIBIT ON-OFF Control**

By setting INHIBIT pin TTL-High, the Device switches off the Output Current and Voltage. Device is ON when INHIBIT pin is set Low. Since INHIBIT pin is internally pulled down, it can be left floating in case Inhibit function is not utilized.

### **Overtemperature protection**

A temperature detector internally monitors the power element junction temperature. The Device goes OFF when approx. 175°C are reached, returning to ON mode when back to approx. 40°C. Combined with other protection blocks, it protects the Device from destructive junction temperature excursions in all load conditions. It is worth noting that when internal temperature detector reaches 175°C, the active power element can be at 225°C: Extensive operation under these conditions far exceeds Maximum Operation Ratings and Device reliability cannot be granted.

### **Overcurrent protection**

An internal non-fold back Short-Circuit limitation is set with  $I_{SHORT} > 3.8A$  (V<sub>O</sub> is 0V). This value can be downwards modified by an external resistor connected between  $I_{SC}$  pin and V<sub>I</sub> pin, with a typical value range of  $10k\Omega$  to  $200k\Omega$ . To keep excellent V<sub>O</sub> regulation, it is necessary to set  $I_{SHORT}$  1.6 times greater than the maximum desired application  $I_O$ . When  $I_O$  reaches  $I_{SHORT} - 300mA$ , the current limiter overrules Regulation and V<sub>O</sub> starts to drop and OCM flag is risen. When no current limit adjustment is required,  $I_{SC}$  pin must be left unbiased (as it is in 3pin packages).

### OCM pin

Goes Low when current limit starts to be active, otherwise  $V_{OCM} = V_I$ . It is buffered and can sink 10mA. OCM pin is internally pulled-up by a 5 k $\Omega$  resistor.

Alternates to: RHFL4913 is recommended to replace all Industry Positive Regulators due to its exceptional Radiation performance. To replace 3-terminal Industry devices, use RHFL4913 Fixed Voltage versions.



### APPLICATION INFORMATION

Adjusting Output Voltage: R2 resistor must be connected between V<sub>O</sub> and ADJ. R1 resistor must be connected between ADJ and Ground. Resistor values can be derived from the following formula:

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V_0 = V_{ADJ} x (R1+R2) / R1
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V<sub>ADJ</sub> is 1.22V, controlled by the internal temperature compensated Band Gap block.

The minimum Output voltage is therefore 1.22V and minimum Input voltage is 3V.

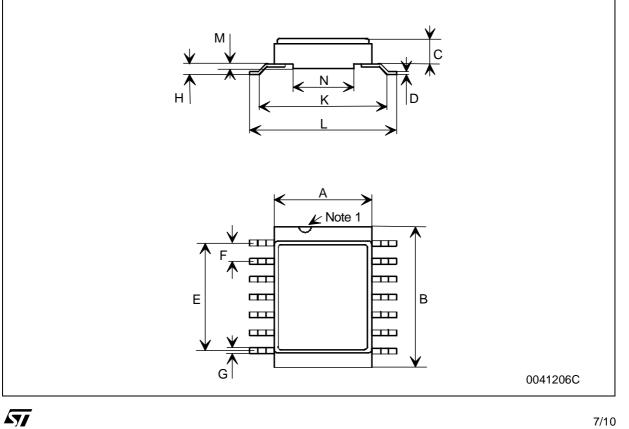
The RHFL4913 Adjustable is functional as soon as  $V_I - V_O$  voltage difference is slightly above the power element saturation voltage. The Adjust pin to Ground resistor must not be bigger than 10k $\Omega$  to keep the output feed-back error below 0.2%. A minimum 0.5mA  $I_O$  must be set to ensure perfect "no-load" regulation. It is advisable to dissipate this current into the divider bridge resistor. All available  $V_I$  pins shall always be externally interconnected, same thing for all available  $V_O$  pins, otherwise Device stability and reliability cannot be granted. The INHIBIT function switches off the output current in an electronic way, that is very quickly. According to Lenz's Law, external circuitry reacts with Ldl/dt terms which can be of high amplitude in case somewhere an inductance exists. Large transient voltage would develop on both Device terminals. It is advisable to protect the Device with Schottky diodes preventing negative voltage excursions. In the worst case, a 14V Zener diode shall protect the Device Input. The Device has been designed for high stability and low drop out operation: Minimum 1µF input and output tantalum capacitors are therefore mandatory. Capacitor ESR range is from 0.5  $\Omega$  to over 20  $\Omega$ . Such range turns out to be useful when ESR increases at low temperature. When large transient currents are expected, larger value capacitors are necessary.

In case of high current operation with expected short-circuit events, caution must be considered relatively to capacitors. They must be connected as close as possible to device terminals. As some tantalum capacitors may permanently fail when submitted to high charge-up surge currents, it is recommended to decouple them with 470nF polyester capacitors.

Being RHFL4913 Adjust manufactured with very high speed bipolar technology (6GHz f<sub>T</sub> transistors), the PCB layout must be performed with exceptional care, very low inductance, low mutually coupling lines, otherwise high frequency parasitic signals may be picked-up by the Device resulting into system self-oscillation. The benefit is an SVR performance extended to far higher frequencies.

DIM.		mm.		inch		
Dini.	MIN.	ТҮР	MAX.	MIN.	TYP.	MAX.
А		6.90			0.272	
В		9.95			0.392	
С	1.49		1.95	0.059		0.077
D	0.102	0.127	0.152	0.004	0.005	0.006
Е	7.5		7.75	0.295		0.305
F		1.27			0.050	
G		0.43			0.017	
Н	0.60		0.90	0.024		0.035
К		9			0.354	
L	10.00		10.65	0.394		0.419
М		0.38			0.015	
N		4.31			0.170	

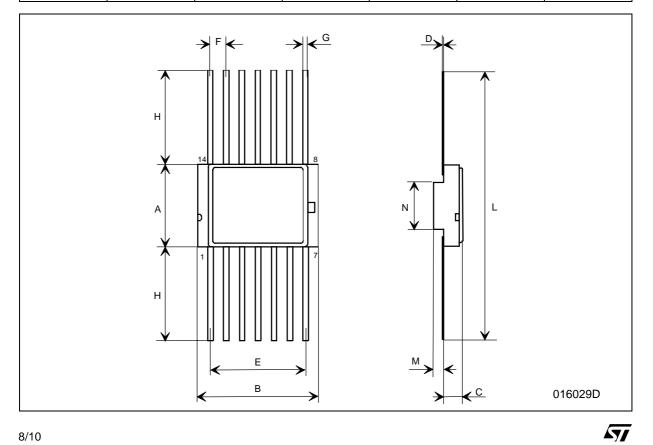




### **RHFL4913 ADJUSTABLE VERSION**

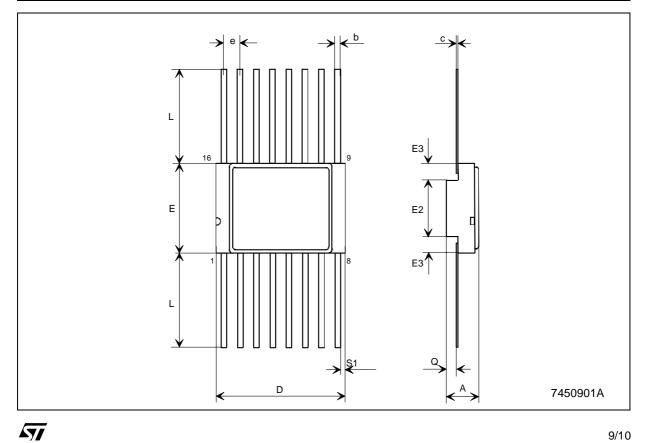
DIM.		mm.			inch			
Dim.	MIN.	ТҮР	MAX.	MIN.	TYP.	MAX.		
А		6.9			0.272			
В		9.95			0.392			
С	1.49		1.95	0.059		0.077		
D		0.127			0.005			
Е		7.62			0.300			
F		1.27			0.050			
G		0.43			0.017			
Н		6.0			0.236			
L	18.75		22.0	0.738		0.866		
М		0.38			0.015			
Ν		4.31			0.170			





DIM.		mm.			inch			
DIN.	MIN.	ТҮР	MAX.	MIN.	TYP.	MAX.		
А	2.16		2.72	0.085		0.107		
b		0.43			0.017			
С		0.13			0.005			
D		9.91			0.390			
Е		6.91			0.272			
E2		4.32			0.170			
E3	0.76			0.030				
е		1.27			0.050			
L		6.72			0.265			
Q	0.66		1.14	0.026		0.045		
S1	0.13			0.005				





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